Abstract — In this paper an approach based on an evolutionary algorithm to design combinational logic circuits with minimum number of Reed Muller units is suggested. Since replication of the same unit reduces the implementation cost of VLSI systems, a single control line Reed Muller universal Logic module (RM ULM) alone is used for the design. Any Boolean function can be realized with this method using any optimization algorithm. Here Genetic Algorithm (GA) is used as the optimization tool. A modification has been made on Davio decomposition technique and it has been observed that the circuits evolved are of lesser complexity and are superior to the circuits in traditional method in terms of power, area and delay.

Index Terms— Combinational logic circuits, Genetic Algorithm, Reed Muller ULM, Davio decomposition technique

I. INTRODUCTION

USE of integrated circuits in high performance computing, telecommunications, consumer electronics etc. is growing at a very high rate with implications in cost effective design. For optimal design of electronic circuits, evolutionary design is a viable alternative. Evolutionary Algorithm is the most important feature in evolvable hardware applications. The algorithm needs to generate optimal circuits, which in turn have to be implemented on a programmable device.

In the design of digital circuits, arriving at the minimal function is of great significance. Using an optimized design the complexity of the circuit can be reduced, thereby reducing the power consumption and cost. Major design criteria involve the delay of the circuit and the area of the chip which directly determines the manufacturing cost. Also power dissipation plays a major role on the packing and cooling cost of the system involved.

Conventional methods like Karnaugh map and Quine McCluskey do not support the use of XOR, XNOR, or any of the basic building blocks like multiplexers or Reed Muller Logic Modules. Any Boolean expression can be realized using universal logic modules like 2-1 multiplexer or 2-1 Reed Muller Universal Logic Modules (RM ULM). Here, 2-1 RM ULM is used as the basic building block. Repeated use of the same unit reduces the manufacturing cost in VLSI implementation [4]-[5].

In comparison to the conventional Boolean logic XOR/AND based Reed Muller logic can be used to express any logic function, and has superior performance on circuits for arithmetic operations, parity check, telecommunication etc.[6]

While implementing a Boolean function, sub functions are useful. To build a hardware circuit for a function with n variables, the following identity is used.

\[ F = F'(A_j)' \text{xor}(F''A_j) \]  
(1)

where \(F'\) and \(F''\) are functions of n-1 variables.

Equation (1) represents Davio decomposition which is valid for realization of functions using Reed Muller units. This identity helps to reduce the original design problem to two simpler problems. Recursively applying such decomposition with every variable allows the process of synthesis to be reduced further to either literals or constants [7].

In this paper, GA based design of logic circuits using single control line RM ULM is proposed. The logic symbol of a 2-1 RM ULM is shown in fig. 1. Its behavior is described as

\[ F= a \oplus b.c \]  
(2)

Fig.1. Logic symbol of a 1-control line RM ULM

By standard implementation technique a function with n variables can be realized using 2^n-1 units in n levels. Any technique which involves lesser number of modules or lesser number of levels is considered as an improvement in the design.
II. RELATED WORK

A Genetic Algorithm based design of combinational logic circuits using minimum number of gates was proposed in [8]. Some of the circuits evolved were 2-1 multiplexer, 4 bit parity checker etc. The drawbacks were the problem of scalability and the time for convergence. In [9] these drawbacks were taken into account by using 2-1 multiplexer as the building block. In [5], design of combinational logic circuits using 2-1 multiplexer and Genetic programming was implemented. Only 4 variable parity circuits could be designed using this method which was a drawback. In [10], authors emphasized the use of RM circuits in certain applications. In [11], they proposed two different algorithms for the design of digital circuits using RM ULM, but it was not based on evolutionary technique. Boolean functions up to 4 variables were tested by them. In [12], authors realized an example of a combinational circuit using 2 variable Reed Muller Binary Decision Diagram and implemented using RM ULM.

In this paper an approach based on an evolutionary algorithm to design combinational logic circuits with minimum number of Reed Muller units is suggested.

The paper is structured as follows. Section III defines the problem, section IV discusses the formulation of algorithm for the evolutionary design, section V gives the implementation of logic circuits using the proposed method and section VI deals with the results and comparison with various methods and section VII concludes the paper.

III. PROBLEM DEFINITION

The problem is to evolve combinational logic circuits using GA with 2-1 RM ULM as the only building block. The goal is to evolve fully functional and optimal circuits, with minimum number of units. Once the fully functional circuit is obtained a check on redundancy is made so that idle units if any can be eliminated.

IV. ALGORITHM FORMULATION

In this work single output combinational logic circuits specified by a truth table is considered. Any combinational digital circuit can be realized using a cascade of RM ULM units.

To implement a function of \( n \) variables, conventional method needs \( n \) levels and \( 2^n-1 \) units. eg., For a 3 variable function, the circuit needs 7 units and 3 levels. There will be 4 units in the first level (bottom most level), 2 units in the second level and 1 unit in the third level. For an optimal circuit, all the modules are not needed [3].

The functional description of the circuit to be evolved is encoded into chromosomes. Chromosomes are generated at random which holds the particulars regarding the presence of a unit, corresponding inputs, control signals and outputs as genes.

Fig. 2 shows the typical representation of a 10 bit chromosome for a unit (RM ULM) in the first level (bottommost layer) of the tree. Bit X1 verifies the presence or absence of an RM ULM and the combination of bits X2 to X6 represents the possible input signals to the RM ULM. The control signal is being determined by the combination of bits X7 to X9. X1=0 indicates the absence of the module.1\(^{st}\)\text{x2}\text{x3}\text{x4}\text{x5}\text{x6}”=“000000”, then the input to the unit is considered as ‘0’, 0’ and so on. Also the combination X7X8X9=“000” implies that the control input is ‘a’. The string “001” corresponds to ‘b’ and so on. If the randomly generated string is “1110110101”, it implies that an RM ULM exists with c and a’ as inputs and control signal as c’.

V. PROPOSED METHOD

Every Boolean function can be expressed in the form of Reed-Muller (RM) expression using AND and XOR operators.

Basic theorems involved are
\[
\begin{align*}
x \bigoplus x &= 0; \quad x \bigoplus 1 = x'; \\
x \bigoplus x' &= 1; \quad x \bigoplus 0 = x;
\end{align*}
\]

Conventional design of combinational logic circuits which is based on Davio decomposition technique permits the use of only same control signals to all the RM blocks in a level and the number of levels needed is equal to the number of variables involved in the function. A modification is being proposed so that circuits are evolved with minimum number of units and/or levels. The control signals need not be the same for all the units in a particular level. They are also generated at random and optimal solution is being achieved by using Genetic Algorithm.

In this paper 0’s, 1’s, variables and their complements are given as inputs to units in the first (bottom most) level. In addition to the above inputs, output of immediate preceding level can also be given as inputs to the units in next level. By conventional Davio decomposition technique, control variables in a particular level are fixed and no flexibility has been provided in giving the control signals. e.g., Control signal in the bottom most layer can be either ‘a’ alone, ‘b’

\[
\text{Fig.2 Chromosomal representation of a RM ULM in the first level}
\]
alone or ‘c’ alone. But in the proposed method control signals are generated at random and need not be the same in a particular level. Circuits have been evolved for 2, 3, 4 and 5 variable functions. Fitness was evaluated based on the fitness function mentioned earlier. Also redundant/idle units are eliminated so that circuits with minimum possible units are generated.

A. Experiments

The program was applied to several functions from the references including some benchmark functions and comparison has been made in terms of the number of units and levels for the generated circuits.

Example 1. Three bit odd parity checker circuit

\[ F(a,b,c) = \sum \oplus (1,2,4,7) \]

The circuit generated for the 3 bit odd parity checker function is shown in fig. 3. It can be seen that the circuit needs only 2 units as compared to 7 units in conventional method. Also the number of levels which decides the delay is reduced from three to two.

Example 2. \[ F(a,b,c) = \sum \oplus (3,5,6) \]

The evolved circuit for this function is shown in fig. 4. The circuit needs only 3 units, but the conventional method by Davio decomposition technique needs 7 units, thus saving 4 units.

Example 3. \[ F(a,b,c,d) = (c+(d \oplus a))' + (b \oplus (a'd')) \]

is a standard benchmark function with 4 inputs and one output.

The circuit generated for this function is shown in fig. 5, which has only 4 units, thus saving 11 units as compared with standard implementation.

Example 4. Four bit odd parity checker

\[ F(a,b,c,d) = \sum \oplus (1,2,4,7,8,11,13,14) \]

The evolved circuit shown in fig. 6 for the function for odd parity has only 3 units and 3 levels, whereas the circuit by conventional method needs 15 units and 4 levels. It is very difficult to evolve circuits for odd parity functions using multiplexers [5]. Here the circuit converges very fast as it involves XOR operations.

Example 5. The next circuit is also a standard benchmark function of 5 variables -xor5.pla.

The evolved circuit shown in fig. 7 has only 4 units and 4 levels. Conventional method needs 31 units in 5 levels. Here the number of levels also got reduced so that the delay involved is also reduced.
Example 6. \( F(a,b,c,d) = \sum \oplus (1,2,3,5,7,8,12) \)

This function is an example taken from [12] where the synthesis was made using 2VRMBDD and GA, which required 6 units and 4 levels. In [3] also the circuit was realized using 6 units and 4 levels. With the proposed method, the circuit is implemented using 4 units and 4 levels as shown in fig.8. Thus a saving of 11 units with respect to the traditional method and 2 units in comparison to [12] has been achieved.

VI. RESULTS

The parameters selected for GA were, crossover rate=0.7, mutation rate=0.3. Population size depends on the complexity of the circuit. As the proposed method is using only RM ULM, circuits with XOR operations converge much faster and the size of populations can be made less. Roulette wheel selection technique has been used for selecting the individuals for crossover. The simulation was done in MATLAB R2012a and analysis of power, area and delay of the evolved circuits have been done by using Synopsys Design Compiler. The program was run on Intel i5duo processor (4.00 GHz).

Each result from GA is taken after running the program 15 to 20 times in order to ensure the quality of the circuit.

Fig.9 and fig.10 show the power consumed and the delay involved by the evolved circuits by standard implementation, method specified in [3] and the proposed method respectively.

Table I shows the comparison of results of the proposed method in terms of number of modules for the above functions with standard implementation technique and the methods proposed in [3]. Method III in [3] is found to be more efficient as compared to other methods. Hence comparison has been made only with standard method and method III of [3]. It is evident that there is a significant reduction in the hardware as the number of units is less.
Table II depicts the % reduction in area involved in the proposed method.

### TABLE I
COMPARISON OF NUMBER OF UNITS AND LEVELS

<table>
<thead>
<tr>
<th></th>
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</thead>
<tbody>
<tr>
<td></td>
<td>No: of units</td>
<td>No: of units</td>
<td>No. of Levels</td>
</tr>
<tr>
<td>1</td>
<td>2</td>
<td>3</td>
<td>2</td>
</tr>
<tr>
<td>2</td>
<td>3</td>
<td>7</td>
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</tr>
<tr>
<td>4</td>
<td>5</td>
<td>31</td>
<td>5</td>
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</table>

### TABLE II
COMPARISON OF AREA INVOLVED FOR CIRCUITS IN EACH METHOD

<table>
<thead>
<tr>
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<th></th>
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</thead>
<tbody>
<tr>
<td></td>
<td>Total Area</td>
<td>Total Area</td>
<td>% Saving</td>
<td>Total Area</td>
</tr>
<tr>
<td>1</td>
<td>2</td>
<td>59.834</td>
<td>38.420</td>
<td>35.78</td>
</tr>
<tr>
<td>2</td>
<td>3</td>
<td>138.031</td>
<td>59.210</td>
<td>57.11</td>
</tr>
<tr>
<td>3</td>
<td>3</td>
<td>38.031</td>
<td>98.589</td>
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</tr>
<tr>
<td>4</td>
<td>4</td>
<td>299.491</td>
<td>78.88</td>
<td>73.67</td>
</tr>
<tr>
<td>5</td>
<td>5</td>
<td>604.142</td>
<td>98.341</td>
<td>83.7</td>
</tr>
</tbody>
</table>

The power consumption and number of units for the circuits in the proposed method and the method in [12] have been analyzed and summarized in tables III and IV respectively.

### TABLE III
COMPARISON OF POWER CONSUMPTION

<table>
<thead>
<tr>
<th>Boolean Function</th>
<th>Standard implementation</th>
<th>Method in [12]</th>
<th>Proposed method</th>
</tr>
</thead>
<tbody>
<tr>
<td>4 input function F(a,b,c,d)=∑(1,2,3,5,7,8,12)</td>
<td>71.535μw</td>
<td>28.86 μw</td>
<td>15.766 μw</td>
</tr>
</tbody>
</table>

### TABLE IV
COMPARISON OF AREA

<table>
<thead>
<tr>
<th>Boolean Function</th>
<th>Standard implementation</th>
<th>Method in [12]</th>
<th>Proposed method</th>
</tr>
</thead>
<tbody>
<tr>
<td>4 input function F(a,b,c,d)=∑(1,2,3,5,7,8,12)</td>
<td>299.491</td>
<td>118.29</td>
<td>78.88</td>
</tr>
</tbody>
</table>

Tables III and IV reveal that the area and power consumption for the circuits in the proposed method reduces significantly as compared to the traditional method and the method proposed in [12].

### VII. CONCLUSION

By evolutionary methods combinational logic circuits have been designed and synthesized using 2-1 RM ULM as the basic building block. On synthesis, it was found that the circuits evolved by this approach are superior to the design by traditional methods. We have used VHDL to describe the developed design and the Synopsys design compiler to synthesize the design. Number of modules, area, delay and power consumption of the evolved circuits were reduced significantly, which ensures better performance.

### REFERENCES


[8] Cecilia Reis, J. A. Tenreiro Machado, and J. Boaventura Cunha,


